FSBS3CH60 Motion SPM[®] 3 Series



SEMICONDUCTOR® FSBS3CH60 Motion SPM® 3 Series

Features

UL Certified No.E209204(SPM27-BA package)

 600 V-3 A 3-Phase IGBT Inverter Bridge Including Control ICs for Gate Driving and Protection

 Three Separate Negative DC-link Terminals for Inverter Current Sensing Applications

· Single-Grounded Power Supply for Built-In HVICs

• Isolation Rating of 2500 Vrms/min.

Motion Control -

MotorFeatures

 \bullet Very Low Leakage Current by Using Ceramic Substrate Sense-IGBT $^{\textcircled{B}}$

Home

Appliance/Industrial

March 2013

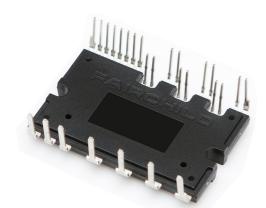
General Description

FSBS3CH60 is a Motion SPM 3 series That Fairchild Has Developed to Provide a Very Compact and High Performance Inverter Solution for AC Motor Drives in Low-Power Applications Such as Air Conditioners and Washing Machines. It Combines Optimized Circuit Protections and Drives Matched to Low-Loss IGBTs. The System Reliability is Further Enhanced by The Integrated Under-Voltage Lock-Out and Over-Current Protection. The High Speed Built-in HVIC Provides Optocoupler-Less Single-Supply IGBT Gate Driving Capability That Further Reduces The Overall Size of The Inverter System. Each Phase Leg Current of The Inverter Can be Monitored Thanks to Three Separate Negative DC Terminals..

RoHS compliant Applications

Resource

• AN-9035: Motion SPM® 3 Series Ver.2 User's Guide





Package Marking and Ordering Information

Device Marking	Device	Package	Reel Size	Packing Type	Quantity
FSBS3CH60	FSBS3CH60	SPMBA-027	-	RAIL	10

Integrated Power Functions

• 600 V-3 A IGBT Inverter for Three-Phase DC/AC Power Conversion (Please Refer to Figure 3)

Integrated Drive, Protection and System Control Functions

- For Inverter High-Side IGBTs: Gate Drive Circuit, High Voltage Isolated High-Speed Level Shifting
 Control Circuit Under-Voltage (UV) Protection
 Note) Available Bootstrap Circuit Example is Given in Figures 10 and 11.
- For Inverter Low-Side IGBTs: Gate Drive Circuit, Short Circuit Protection (SC) Control Supply Circuit Under-Voltage (UV) Protection
- Fault Sgnaling: Corresponding to a UV Fault (Low-Side Supply)
- Input interface: Active High Interface, Can Work with 3.3/5 V Logic

Pin Configuration

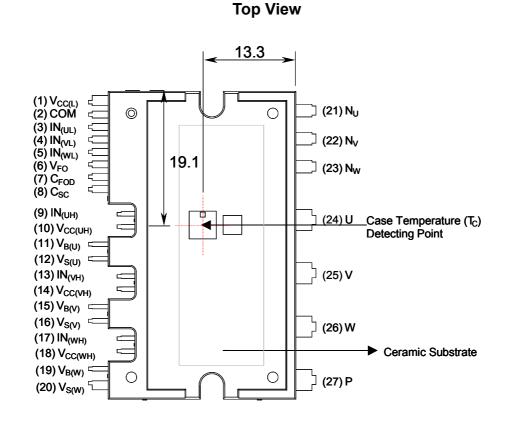
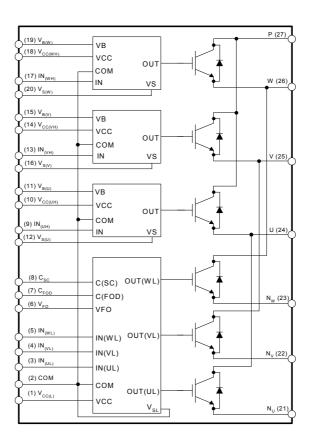


Figure 2.

Pin Number	Pin Name	Pin Description	
1	V _{CC(L)}	Low-side Common Bias Voltage for IC and IGBTs Driving	
2	COM	Common Supply Ground	
3	IN _(UL)	Signal Input for Low-side U Phase	
4	IN _(VL)	Signal Input for Low-side V Phase	
5	IN _(WL)	Signal Input for Low-side W Phase	
6	V _{FO}	Fault Output	
7	C _{FOD}	Capacitor for Fault Output Duration Time Selection	
8	C _{SC}	Capacitor (Low-pass Filter) for Short-Current Detection Input	
9	IN _(UH)	Signal Input for High-side U Phase	
10	V _{CC(UH)}	High-side Bias Voltage for U Phase IC	
11	V _{B(U)}	High-side Bias Voltage for U Phase IGBT Driving	
12	V _{S(U)}	High-side Bias Voltage Ground for U Phase IGBT Driving	
13	IN _(VH)	Signal Input for High-side V Phase	
14	V _{CC(VH)}	High-side Bias Voltage for V Phase IC	
15	V _{B(V)}	High-side Bias Voltage for V Phase IGBT Driving	
16	V _{S(V)}	High-side Bias Voltage Ground for V Phase IGBT Driving	
17	IN _(WH)	Signal Input for High-side W Phase	
18	V _{CC(WH)}	High-side Bias Voltage for W Phase IC	
19	V _{B(W)}	High-side Bias Voltage for W Phase IGBT Driving	
20	V _{S(W)}	High-side Bias Voltage Ground for W Phase IGBT Driving	
21	NU	Negative DC–Link Input for U Phase	
22	N _V	Negative DC–Link Input for V Phase	
23	N _W	Negative DC–Link Input for W Phase	
24	U	Output for U Phase	
25	V	Output for V Phase	
26	W	Output for W Phase	_
27	Р	Positive DC-Link Input	

FSBS3CH60 Motion SPM[®] 3 Series

Internal Equivalent Circuit and Input/Output Pins



Note:

1. Inverter low-side is composed of three IGBTs, freewheeling diodes for each IGBT and one control IC. It has gate drive and protection functions.

2. Inverter power side is composed of four inverter dc-link input terminals and three inverter output terminals.

3. Inverter high-side is composed of three IGBTs, freewheeling diodes and three drive ICs for each IGBT.



Absolute Maximum Ratings (T_J = 25°C, Unless Otherwise Specified)

Inverter Part

Symbol	Parameter	Conditions	Rating	Unit
V _{PN}	Supply Voltage	Applied between P- N_U , N_V , N_W	450	V
V _{PN(Surge)}	Supply Voltage (Surge)	Applied between P- N_U , N_V , N_W	500	V
V _{CES}	Collector-emitter Voltage		600	V
± I _C	Each IGBT Collector Current	$T_{\rm C} = 25^{\circ}{\rm C}$	3	A
± I _{CP}	Each IGBT Collector Current (Peak)	T _C = 25°C, Under 1ms Pulse Width	6	A
P _C	Collector Dissipation	T _C = 25°C per One Chip	23	W
ТJ	Operating Junction Temperature	(Note 1)	-20 ~ 125	°C

Note:

The maximum junction temperature rating of the power chips integrated within the SPM is 150 °C(@T_C \leq 100°C). However, to insure safe operation of the SPM, the average junction temperature should be limited to T_{J(ave)} \leq 125°C (@T_C \leq 100°C)

Control Part

Symbol	Parameter	Conditions	Rating	Unit
V _{CC}	Control Supply Voltage	Applied between $V_{CC(UH)}$, $V_{CC(VH)}$, $V_{CC(WH)}$, $V_{CC(L)}$ - COM	20	V
V _{BS}	High-side Control Bias Voltage	Applied between V _{B(U)} - V _{S(U)} , V _{B(V)} - V _{S(V)} , V _{B(W)} - V _{S(W)}	20	V
V _{IN}	Input Signal Voltage	Applied between $\rm IN_{(UH)},~\rm IN_{(VH)},~\rm IN_{(WH)},~\rm IN_{(UL)},~\rm IN_{(VL)},~\rm IN_{(VL)},~\rm IN_{(WL)},~\rm IN_{(WL)},~$	-0.3~17	V
V _{FO}	Fault Output Supply Voltage	Applied between V _{FO} - COM	-0.3~V _{CC} +0.3	V
I _{FO}	Fault Output Current	Sink Current at V _{FO} Pin	5	mA
V _{SC}	Current Sensing Input Voltage	Applied between C _{SC} - COM	-0.3~V _{CC} +0.3	V

Total System

Symbol	Parameter	Conditions	Rating	Unit
V _{PN(PROT)}	Self Protection Supply Voltage Limit (Short Circuit Protection Capability)	$V_{CC} = V_{BS} = 13.5 \sim 16.5 V$ T _J = 125°C, Non-repetitive, less than 2µs	400	V
т _с	Module Case Operation Temperature	-20°C \leq T _J \leq 125°C, See Figure 2	-20 ~ 100	°C
T _{STG}	Storage Temperature		-40 ~ 125	°C
V _{ISO}	Isolation Voltage	60Hz, Sinusoidal, AC 1 minute, Connection Pins to ceramic substrate	2500	V _{rms}

Thermal Resistance

Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
R _{th(j-c)Q}	Junction to Case Thermal	Inverter IGBT part (per 1/6 module)	-	-	4.1	°C/W
R _{th(j-c)F}	Resistance	Inverter FWD part (per 1/6 module)	-	-	5.9	°C/W

Note:

2. For the measurement point of case temperature(T_C), please refer to Figure 2.

Electrical Characteristics (T_J = 25°C, Unless Otherwise Specified)

Inverter Part

Sy	ymbol	Parameter	Cond	itions	Min.	Тур.	Max.	Unit
V	CE(SAT)	Collector-Emitter Saturation Voltage	V _{CC} = V _{BS} = 15 V V _{IN} = 5 V	I _C = 3 A, T _J = 25°C	-	-	2.3	V
	V _F	FWD Forward Voltage	$V_{IN} = 0 V$	$I_{\rm C}$ = 3 A, T _J = 25°C	-	-	2.1	V
HS	t _{ON}	Switching Times	V_{PN} = 300 V, V_{CC} = V_{E}	_{3S} = 15 V	-	0.39	-	μs
	t _{C(ON)}		$I_C = 3 A$ $V_{IN} = 0 V \leftrightarrow 5 V$, Induc	tive Load	-	0.19	-	μs
	t _{OFF}		(Note 3) (Note 3)		-	0.56	-	μs
	t _{C(OFF)}				-	0.19	-	μs
	t _{rr}				-	0.10	-	μs
LS	t _{ON}		V _{PN} = 300 V, V _{CC} = V _E	_{3S} = 15 V	-	0.57	-	μs
	t _{C(ON)}		$I_C = 3 A$ $V_{IN} = 0 V \leftrightarrow 5 V$, Induc	tive Load	-	0.24	-	μs
	t _{OFF}		(Note 3) (Note 3)		-	0.62	-	μs
	t _{C(OFF)}				-	0.20	-	μs
	t _{rr}]			-	0.10	-	μs
	I _{CES}	Collector-Emitter Leakage Current	V _{CE} = V _{CES}		-	-	250	μΑ

Note:

3. t_{ON} and t_{OFF} include the propagation delay time of the internal drive IC. $t_{C(OFF)}$ are the switching time of IGBT itself under the given gate driving condition internally. For the detailed information, please see Figure 4.

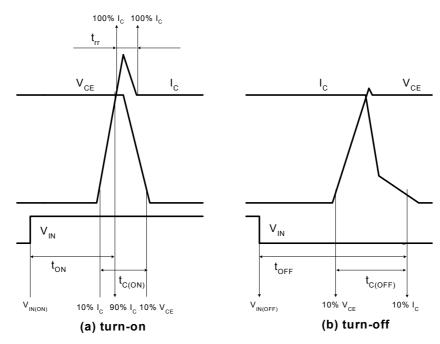


Figure 4. Switching Time Definition

Electrical Characteristics (T_J = 25°C, Unless Otherwise Specified)

Control Part

Symbol	Parameter	Co	nditions	Min.	Тур.	Max.	Unit
I _{QCCL}	Quiescent V _{CC} Supply Current	V _{CC} = 1 5V IN _(UL, VL, WL) = 0 V	V _{CC(L)} - COM	-	-	23	mA
I _{QCCH}		V _{CC} = 15V IN _(UH, VH, WH) = 0 V	$\begin{array}{l} V_{CC(UH)}, V_{CC(VH)}, \\ V_{CC(WH)} - COM \end{array}$	-	-	100	μA
I _{QBS}	Quiescent V _{BS} Supply Current	V _{BS} = 15 V IN _(UH, VH, WH) = 0 V		-	-	500	μA
V _{FOH}	Fault Output Voltage	V _{SC} = 0 V, V _{FO} Circu	V_{SC} = 0 V, V_{FO} Circuit: 4.7 k Ω to 5 V Pull-up		-	-	V
V _{FOL}		V _{SC} = 1 V, V _{FO} Circu	iit: 4.7 kΩ to 5 V Pull-up	-	-	0.8	V
V _{SC(ref)}	Short Circuit Trip Level	V _{CC} = 15 V (Note 4)		0.45	0.5	0.55	V
UV _{CCD}	Supply Circuit Under-	Detection Level		10.7	11.9	13.0	V
UV _{CCR}	Voltage Protection	Reset Level		11.2	12.4	13.2	V
UV _{BSD}		Detection Level		10.1	11.3	12.5	V
UV _{BSR}		Reset Level		10.5	11.7	12.9	V
t _{FOD}	Fault-out Pulse Width	C _{FOD} = 33 nF (Note 5)		1.0	1.8	-	ms
V _{IN(ON)}	ON Threshold Voltage	Applied between IN	Applied between IN _(UH) , IN _(VH) , IN _(WH) , IN _(UL) ,		-	-	V
V _{IN(OFF)}	OFF Threshold Voltage	IN _(VL) , IN _(WL) - COM		-	-	0.8	V

Note:

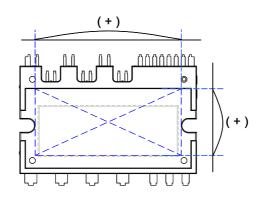
4. Short-circuit current protection is functioning only at the low-sides.

5. The fault-out pulse width t_{FOD} depends on the capacitance value of C_{FOD} according to the following approximate equation : C_{FOD} = 18.3 x 10⁻⁶ x $t_{FOD}[F]$

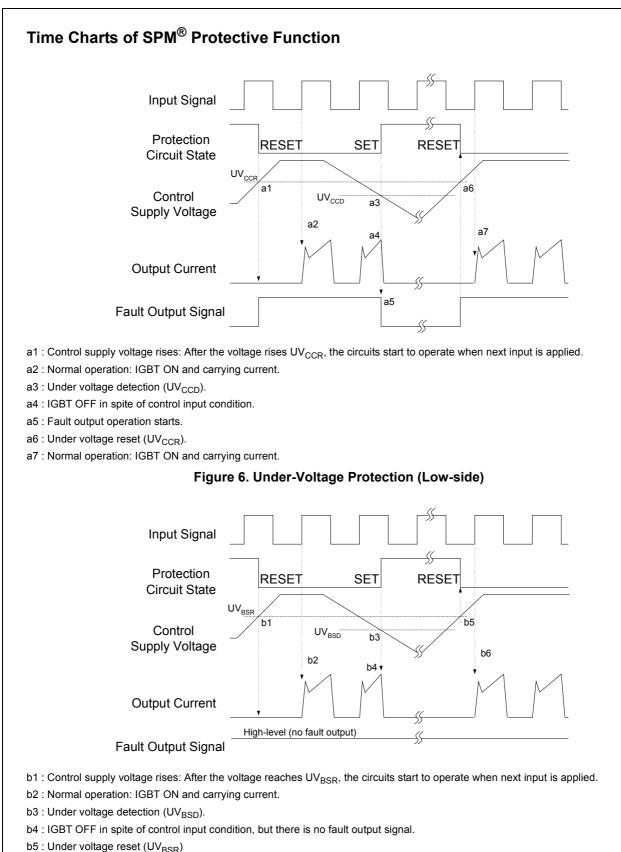
Recommended Operating Conditions

Symbol	Parameter	Conditions	Value			Unit
Symbol	Farameter	Conditions	Min.	Тур.	Max.	Unit
V _{PN}	Supply Voltage	Applied between P - N _U , N _V , N _W	-	300	400	V
V _{CC}	Control Supply Voltage Applied between V _{CC(UH)} , V _{CC(VH)} , V _{CC(WH)} , V _{CC(WH)} , V _{CC(UH)} , V _{CC(WH)} ,		13.5	15	16.5	V
V _{BS}	High-side Bias Voltage	Applied between V_B(U) - V_S(U), V_B(V) - V_S(V), V_B(W) - V_S(W)	13.0	15	18.5	V
DV _{CC} /Dt, DV _{BS} /Dt	Control supply variation		-1	-	1	V/µs
t _{dead}	Blanking Time for Preventing Arm-short	For Each Input Signal	2	-	-	μs
f _{PWM}	PWM Input Signal $-20^{\circ}C \le T_C \le 100^{\circ}C, -20^{\circ}C \le T_J \le 125^{\circ}C$		-	-	20	kHz
V _{SEN}	Voltage for Current Sensing	Applied between N _U , N _V , N _W - COM (Including surge voltage)	-4		4	V

Parameter	6.	nditions		Unit		
Falametei			Min.		Max.	Onit
Mounting Torque	Mounting Screw: - M3	Recommended 0.62N•m	0.51	0.62	0.72	N•m
Device Flatness		Note Figure 5	0	-	+120	μm
Weight			-	15.4	-	g

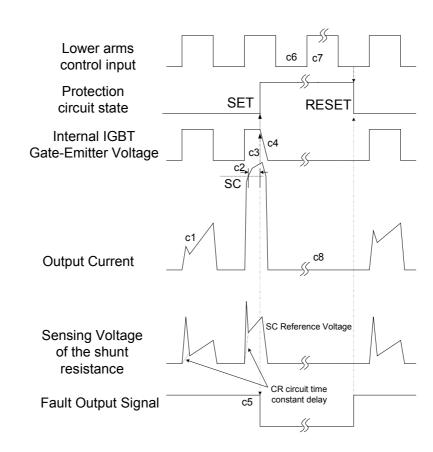






b6 : Normal operation: IGBT ON and carrying current

Figure 7. Under-Voltage Protection (High-side)



(with the external shunt resistance and CR connection)

c1 : Normal operation: IGBT ON and carrying current.

c2 : Short circuit current detection (SC trigger).

c3 : Hard IGBT gate interrupt.

c4 : IGBT turns OFF.

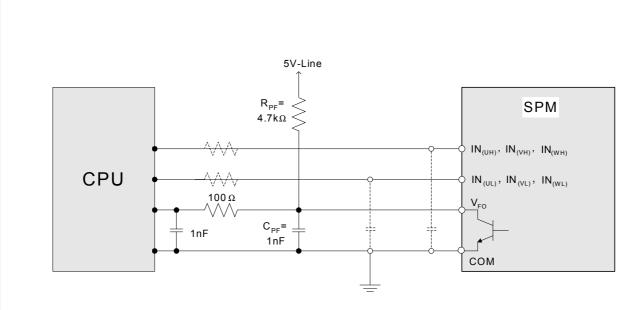
c5 : Fault output timer operation starts: The pulse width of the fault output signal is set by the external capacitor C_{FO} .

c6 : Input "L" : IGBT OFF state.

c7 : Input "H": IGBT ON state, but during the active period of fault output the IGBT doesn't turn ON.

c8 : IGBT OFF state

Figure 8. Short-Circuit Current Protection (Low-side Operation only)

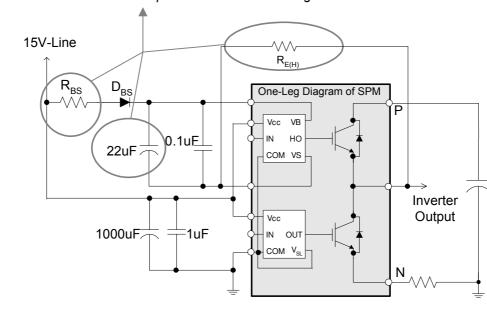


Note:

1. RC coupling at each input (parts shown dotted) might change depending on the PWM control scheme used in the application and the wiring impedance of the application's printed circuit board. The SPM input signal section integrates 3.3kΩ (typ.) pull-down resistor. Therefore, when using an external filtering resistor, please pay attention to the signal voltage drop at input terminal.

2. The logic input is compatible with standard CMOS or LSTTL outputs.

Figure 9. Recommended CPU I/O Interface Circuit



These Values depend on PWM Control Algorithm

Note:

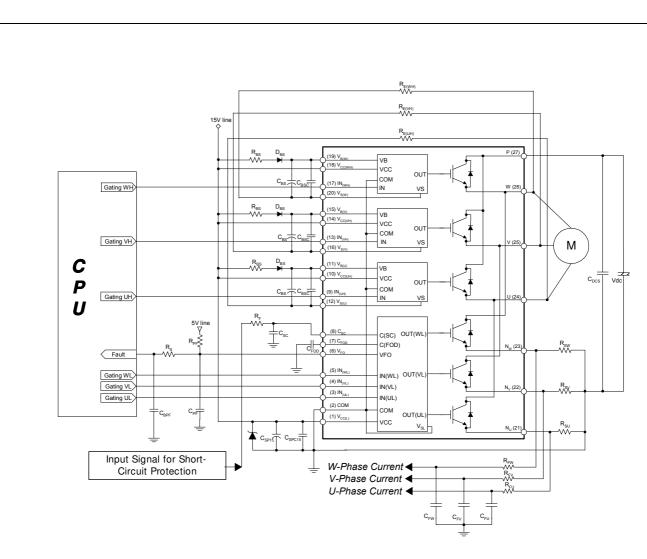
1. It would be recommended that the bootstrap diode, D_{BS}, has soft and fast recovery characteristics.

2. The bootstrap resistor (R_{BS}) should be 3 times greater than R_{E(H)}. The recommended value of R_{E(H)} is 5.6Ω, but it can be increased up to 20Ω (maximum) for a slower dv/dt of high-side.

3. The ceramic capacitor placed between V_{CC}-COM should be over 1uF and mounted as close to the pins of the SPM as possible.

Fig. 10. Recommended Bootstrap Operation Circuit and Parameters

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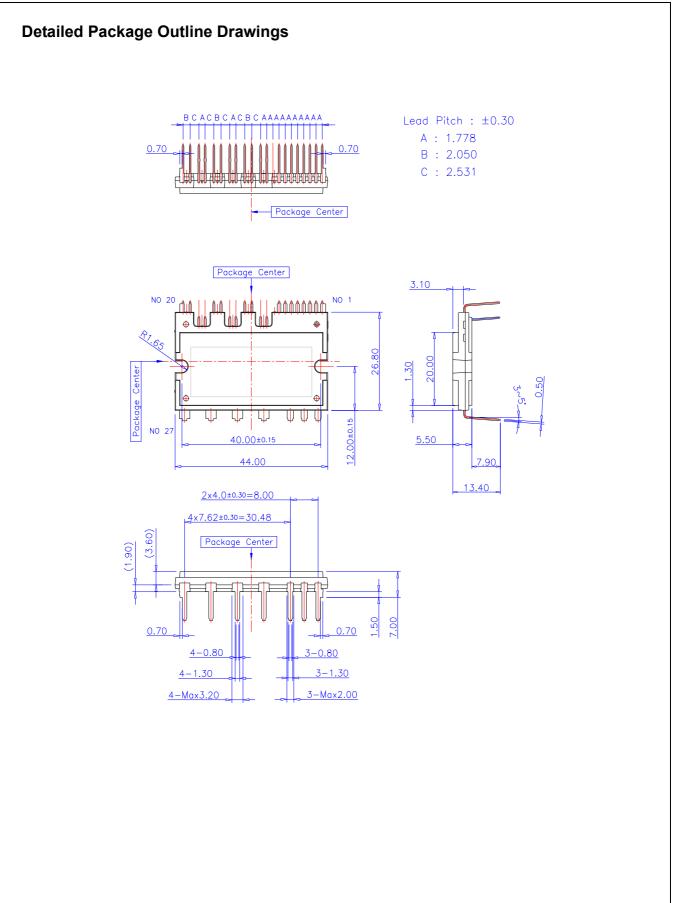


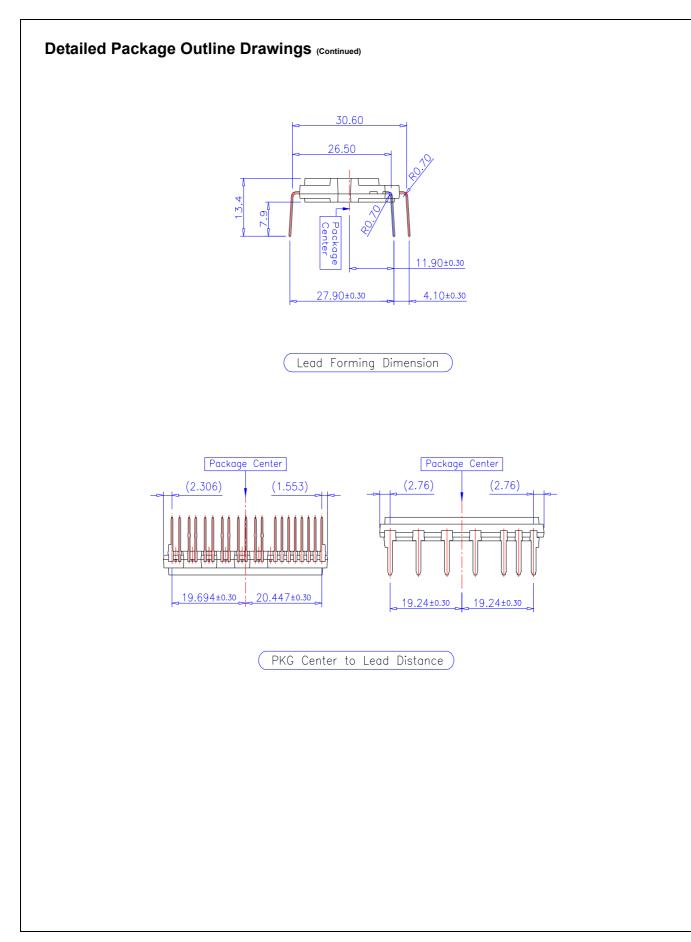
Note:

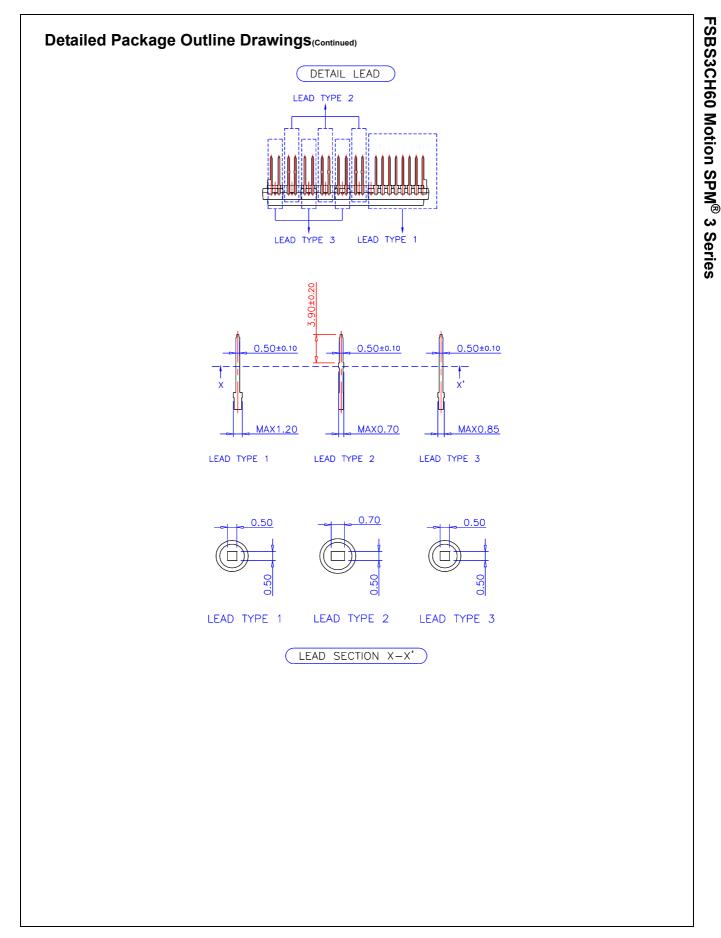
- 1. To avoid malfunction, the wiring of each input should be as short as possible. (less than 2-3cm)
- 2. By virtue of integrating an application specific type HVIC inside the SPM, direct coupling to CPU terminals without any opto-coupler or transformer isolation is possible.
- 3. V_{FO} output is open collector type. This signal line should be pulled up to the positive side of the 5V power supply with approximately 4.7k Ω resistance. Please refer to Figure 9. 4. C_{SP15} of around 7 times larger than bootstrap capacitor C_{BS} is recommended.
- 5. V_{FO} output pulse width should be determined by connecting an external capacitor(C_{FOD}) between C_{FOD} (pin7) and COM(pin2). (Example : if C_{FOD} = 33 nF, then t_{FO} = 1.8ms (typ.)) Please refer to the note 5 for calculation method.
- 6. Input signal is High-Active type. There is a 3.3kΩ resistor inside the IC to pull down each input signal line to GND. When employing RC coupling circuits, set up such RC couple that input signal agree with turm-off/turn-on threshold voltage.
- 7. To prevent errors of the protection function, the wiring around R_F and C_{SC} should be as short as possible.
- 8. In the short-circuit protection circuit, please select the R_FC_{SC} time constant in the range 1.5~2 μ s.
- 9. Each capacitor should be mounted as close to the pins of the SPM as possible.
- 10. To prevent surge destruction, the wiring between the smoothing capacitor and the P&GND pins should be as short as possible. The use of a high frequency non-inductive capacitor of around 0.1~0.22 uF between the P&GND pins is recommended.
- 11. Relays are used at almost every systems of electrical equipments of home appliances. In these cases, there should be sufficient distance between the CPU and the relays.

12. C_{SPC15} should be over 1uF and mounted as close to the pins of the SPM as possible.

Fig. 11. Typical Application Circuit







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